

Optimized Lattice-based 16-level Subcarrier Modulation for IM/DD Systems

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Abstract We present an experimental demonstration of an optimized 16-level lattice-based single-cycle subcarrier modulation for IM/DD systems at 10 Gbps. The new format has 2.5 dB better sensitivity than single-cycle subcarrier 16-QAM and up to 1 dB better sensitivity than baseband 4-PAM at the same bit rate.

Introduction

The increasing popularity of cloud computing and demand for larger datacenters is currently a driving factor for a development of high-speed short-range optical interconnects. Such optical links are usually built with low-cost and high-speed vertical cavity surface emitting lasers (VCSELs). Intensity modulation and direct detection (IM/DD) is used in such systems because of cost constraints. The prevalent modulation format in commercial links is on-off keying (OOK).

Recently, research on more spectrally efficient modulation formats has become very active. Modulation formats such as a single-cycle subcarrier 16-level quadrature amplitude modulation (QAM) and a baseband 4-level pulse amplitude modulation (PAM) have been demonstrated at, respectively, 37 Gbps¹ and 30 Gbps². A big advantage of PAM is its simplicity, allowing real-time and high-speed implantations. Real-time, high-speed implementations of subcarrier QAM are difficult because of its complexity. On the other hand, carrier-less amplitude and phase modulation (CAP), which also uses in-phase and quadrature pulses as basis functions, has been demonstrated experimentally at 40 Gbps in real time³. Transversal filters were used to generate the CAP signal³. Orthogonal multi-pulse modulation with three orthogonal basis functions has also been demonstrated in real-time using the same method⁴.

Most multi-level modulation formats have a disadvantage of reduced sensitivity, compared to OOK. Theoretically, 4.8 dB more optical power is required to double the bit rate by using baseband 4-PAM instead of OOK at the same symbol rate². The sensitivity of modulation schemes is an important consideration because the optical power budget is limited by factors such as limited output power of VCSELs, power

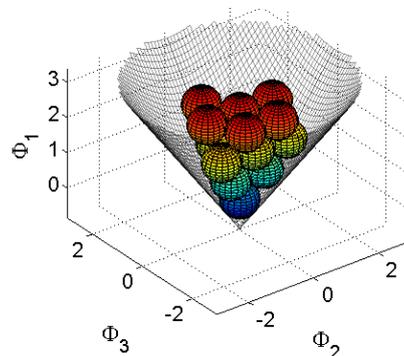


Fig. 1: Constellation diagram of a 16-level single-cycle subcarrier format based on face centered cubic lattice (L_{16}).

handling abilities of detectors, losses, or eye safety constraints.

Three dimensional single-cycle subcarrier modulation formats

The two dimensions (in-phase and quadrature-phase, denoted here Φ_2 and Φ_3) of the single-cycle subcarrier modulation can be complemented with a third orthogonal dimension, represented by a rectangular pulse

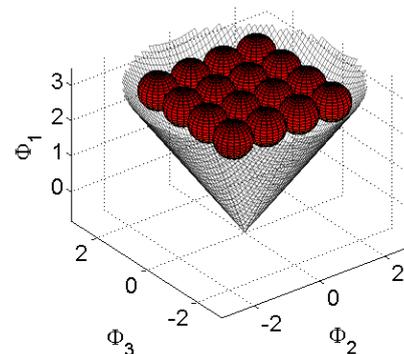


Fig. 2: Subcarrier 16-QAM constellation represented in a three dimensional signal space.

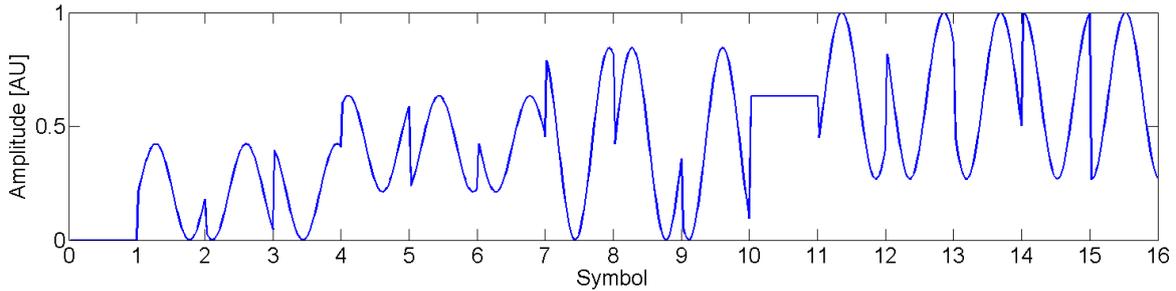


Fig. 3: Symbols of the L_{16} modulation presented in time domain.

of the same duration as the symbol period (denoted Φ_1). It can be thought of as an adaptive bias for the subcarrier symbols. The resultant three-dimensional signal space was analyzed in⁷.

For IM/DD systems with directly modulated lasers, any signal modulating the laser must be non-negative to avoid clipping, and the electrical subcarrier has to be biased to make it non-negative. In the simplest case, the subcarrier signal is shifted by the same amount for all symbols, even when for some symbols a lower bias would suffice. Consequently, some of the optical power does not carry information, because the modulation depth is lower.

The Φ_1 dimension allows for so called “adaptive biasing”, which can improve the sensitivity of the subcarrier modulation⁵ and was demonstrated experimentally⁶ for 8-level single-cycle subcarrier modulation. Modulation formats with sensitivity better than PAM at the same spectral efficiency were proposed in⁸. For example, at the spectral efficiency of 2 bits/s/Hz and at the same bit rate, sensitivity better by up to 1.5 dB than 4-PAM can be achieved with an optimized 16-level subcarrier modulation⁸. The constellation shape is irregular in this case. It is also possible to perform an optimization of lattice-based constellations, which results in more regular shapes. The face-centered cubic (FCC) lattice provides the best packing of symbols in a three dimensional space. A format based on the FCC lattice was proposed in⁸ (denoted as L_{16}), with an asymptotical sensitivity 1 dB better than baseband 4-PAM and 2.5 dB better sensitivity than single-cycle subcarrier 16-QAM. The theoretical performance gains are valid for systems with additive white Gaussian noise (AWGN). The symbols of L_{16} were given in⁸. Using $a = \sqrt{2/3}$, $b = 1/\sqrt{3}$, and $c = \sqrt{3}/6$ the symbols in L_{16} can be defined in the $\{\Phi_1, \Phi_2, \Phi_3\}$ signal space as follows: $\{(0, 0, 0), (a, 0, b), (a, \pm 1/2, -c), (2a, \pm 1/2, c), (2a, 0, -b), (2a, \pm 1, -b), (2a, 0, 2b), (3a, 0, 0), (3a, \pm 1, 0), (3a, \pm 1/2, -3c), (3a, 1/2, 3c)\}$.

An illustration of the L_{16} constellation in three dimensions is presented in Fig. 1, with spheres

to visualize the locations of symbol. The spheres have a diameter of one minimum distance between the symbols. The symbols are located at the centers of the spheres. The cone in Fig. 1 illustrates the admissible signal space defined by the non-negativity constraint, i.e. the part of the signal space which gives symbols with non-negative waveforms⁷. For comparison the subcarrier 16-QAM constellation is shown in the same signal space in Fig.2. Obviously L_{16} utilizes the admissible signal space in a more efficient way. An illustration of an L_{16} symbol waveform in time domain is shown in Fig. 3.

Experimental setup

Three modulation formats with the same spectral efficiency were experimentally compared, L_{16} , baseband 4-PAM and 16-QAM with a single-cycle subcarrier. All formats were compared at 10 Gbps, which means that 4-PAM was operated at 5 Gbaud and L_{16} and subcarrier 16-QAM were operated at 2.5 Gbaud, since the bandwidth of the single-cycle subcarrier signal is twice the symbol rate¹. The electrical bandwidth of all signals was 5 GHz, counted to the first spectral null. The modulation formats were tested in a link consisting of a directly modulated VCSEL operating at a wavelength of 850 nm with a 16 GHz modulation bandwidth, multimode fiber patch-cord and a photoreceiver with a 12 GHz modulation bandwidth, for direct detection. The VCSEL was from the same batch as the one reported in⁹. All modulation formats were generated using an arbitrary waveform generator and after transmission detected and sampled using a real time sampling oscilloscope and further processed off-line on a personal computer to obtain the symbol error rates (SER). Because of the off-line processing the lowest SER values obtainable with reasonable effort were around 10^{-5} . The modulator and demodulator for L_{16} were based on the classical I/Q structure for ordinary QAM and PSK formats, but with an extra branch for the third dimension, which represents only a small increase in complexity over usual subcarrier modulator or

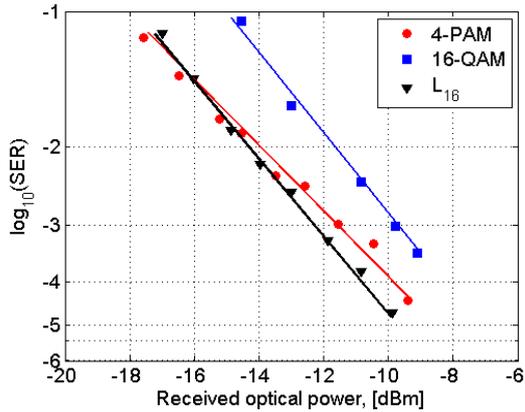


Fig. 4: Experimental bit-error rates for different 2bit/s/Hz formats in an IM/DD VCSEL link

demodulator. In the off-line receiver a maximum likelihood detector was used, since it minimizes the SER for AWGN channels.

Results and discussion

The experimentally obtained SER values for a back to back (BTB) case are presented in Fig. 4. The bit error rate would be dependent on the bit-labeling of the symbols. The optimal labeling for L_{16} remains to be found. An example of a received three-dimensional constellation diagram of L_{16} is illustrated in Fig. 5.

In the BTB configuration, the sensitivity of the L_{16} format is almost 1 dB better than 4-PAM at SER of 10^{-4} and at 2.5 dB than single-cycle subcarrier 16-QAM. This is agreement with the theoretical expectations from⁸.

There are few remarks on the observed results. At SER above 10^{-2} the performance of 4-PAM and L_{16} is similar. In theory the sensitivity gain of L_{16} over 4-PAM is increasing to the asymptotical value of 1 dB for low SER⁸. The experimentally demonstrated sensitivity gain of

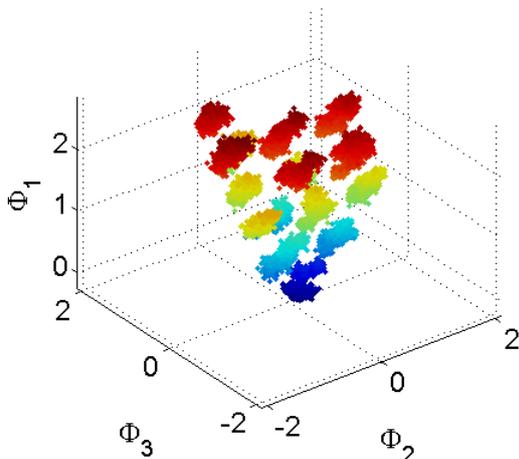


Fig. 5: Constellation diagram of L_{16} at -11 dBm received optical power. The different Φ_1 levels are shown with different colors.

L_{16} over 4-PAM is increasing to almost 1 dB. Finally, although the transmission bit rate of the presented modulation formats is 10 Gbps, for practical data transmission forward error correction (FEC) would have to be used at the observed SER. With 7% FEC overhead the usable bitrate is 9.3 Gbps.

Conclusions

We have experimentally demonstrated an optimized lattice-based 16-level single-subcarrier modulation format with 1 dB better sensitivity than 4-PAM modulation, at 10 Gbps data rates.

Although the presented work was done with off-line processing, it should be possible to implement it in real time, in a manner similar to the CAP modulation³. The somewhat higher complexity of the L_{16} compared to the single-cycle subcarrier 16-QAM, is traded for 2.5 dB sensitivity gain.

Potential applications of the L_{16} format, beyond IM/DD systems with VCSELS and MMF could include for example polymer optical fiber networks or wireless optical transmission with IM/DD.

Acknowledgements

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References

- [1] K. Szczerba et al., Proc. ECOC'10, We.7.B.2 (2010).
- [2] K. Szczerba et al., Opt. Express **19**, B203-B208 (2011).
- [3] J. D. Ingham et al., Proc. OFC'11, OThZ3 (2011).
- [4] J. D. Ingham et al., Proc. ECOC'11, Tu.3.C.2 (2011).
- [5] S. Hranilovic et al., Proc. ISCAS'99, 338–341, (1999).
- [6] K. Szczerba et al., Proc. ECOC'11, We.10.P1.117 (2011).
- [7] S. Hranilovic et al., IEEE Trans. Inf. Theory, **49**, 1385–1399, (2003).
- [8] J. Karout et al., to appear in IEEE Trans. Inf. Theory, available: <http://arxiv.org/abs/1106.2819>.
- [9] P. Westbergh et al. Electron. Lett., **45**, 366–368, (2009).